

PERFORMANCE ANALYSIS OF FIFTH GENERATION WIRELESS COMMUNICATIONS

Submitted in partial fulfillment of the
requirements for
the award of
Bachelor of Engineering degree in Computer Science and Engineering

By

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**INSTITUTE OF SCIENCE AND TECHNOLOGY
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Accredited with Grade "A" by NAAC

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BONAFIDE CERTIFICATE

This is to certify that this Project Report is the bonafide work P.Tarun Teja (38110594), V.Srinivas Reddy (38110617) who carried out the project entitled as "Performance Analysis of Fifth Generation Wireless Communications" under my supervision from to

A.C. SANTHA SHEELA

InternalGuide

(Name in Capital letters with signature)

Head of the Department

Submitted for Viva voce Examination held on _____

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DECLARATION

I P. Tarun Teja, V. Srinivas Reddy hereby declare that the Project Report
entitled **Performance Analysis of Fifth Generation Wireless
Communications**

done by me under the guidance of Dr./Prof./Mr./Ms. A.C. Santha Sheela (Internal) is su

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CANDIDATE

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ABSTRACT

In 4G, the using of Orthogonal Multiple Access techniques (OMA) which does not fulfill the massive increase in Mobile networks and also Increasing Technology like IOT. It was limiting the number of users that can be supported by the number of orthogonal resources available, as it was not fulfilling the exponential growth in Mobile network like low latency, massive connectivity and energy efficiency etc.

So to overcome the problems with OMA in 4G, In 5G NOMA was implemented which overcomes problems to some extent. Like it uses the same resource block. It transmit multiple users' signals from the transmitter to the receiver. By allowing multiple users to share the same resource elements, be it in the time, frequency, space, or code domain. The main advantage of NOMA is it uses power domain technique for transmission of signals.

As NOMA is versatile technique. we can combine it with MIMO to increase throughput as well as to increase spatial multiplexing to increase achievable rates and diversity gain to decrease BER.

Our main aim of this project is to show the performance evaluation of Multi-Input Multi-Output Non-Orthogonal Multiple Access (MIMO-NOMA) for Next Generation Wireless Communication Systems. The efficiency of the MIMO-NOMA technique over Non- Orthogonal Multiple Access (NOMA) and Orthogonal Multiple Access (OMA) is examined by using distinct parameters, in particular, Power Allocation (PA), Bit Error Rate (BER), Outage Probability (OP), and Channel Capacity (CC).

A system model for MIMO is assumed for the two-user scenario. A Rayleigh fading channel is assumed between the Transmitter and the receiver. By using Rayleigh fading channel coefficients, calculated the

achievable rates and SNR equations for all techniques and then compared each parameter with others.

Finally, the simulation results are provided to facilitate the performance analysis of MIMO- NOMA for a next-generation wireless communication system and also demonstrate the accuracy of the developed analytical results.

LIST OF FIGURES:

- Figure 3.1: NOMA transmission in DLmode
- Figure 3.2: NOMA in transmissionsection
- Figure 3.3: Decoding atreceiver
- Figure 3.4: MIMO-OMA Downlink systemmodel
- Figure 3.5: NOMA Downlink systemmodel
- Figure 3.6: MIMO Downlink systemmodel
- Figure 4.1: Power Allocation of MIMO-NOMA, NOMA, andOMA
- Figure 4.2: Channel Capacity of MIMO-NOMA, NOMA, andOMA
- Figure 4.3: Outage Probability of MIMO-NOMA, NOMA, andOMA
- Figure 4.4: Bit Error Rate of MIMO-NOMA, NOMA, andOMA

LIST OF ABBREVIATIONS:

- OMA - Orthogonal MultipleAccess
- NOMA - Non - Orthogonal MultipleAccess
- MIMO-NOMA - Multiple Input Multiple
Output Non - Orthogonal
MultipleAccess
- SIC - Successive InterferenceCancellation
- SNR - Signal to NoiseRatio
- BER - Bit ErrorRate
- PA - PowerAllocation
- CC - ChannelCapacity
- OP - OutageProbability
- BEP - Bit ErrorProbability
- BS - BaseStation

CHAPTER-1

INTRODUCTION

Different multiple access schemes are acquired from First Generation (1G) to Fourth Generation (4G). All the proposed schemes from 1G to 4G are common in one particular subject. That is the signals that are transmitted from the transmitter to the receiver must be orthogonal to each other.

OMA is one of the multiple access schemes which uses orthogonal resource allocation among individual users to avoid intra-cell interference. This technique also helps in the dissociation of respective user data in which time and frequency resources are split for multiple receivers in current broadcasting systems. But the number of users that can be supported through this OMA method is then limited by the number of orthogonal resources available. As 4G, uses the OMA approach, that cannot satisfy high-speed communication or with low latency, etc. The main disadvantages of 4G networks are, they don't accomplish the exponential growth in capacity demands, low latency, massive connectivity, high throughput, and energy efficiency.

A lot of investigations have been done to satisfy various requirements in practical systems in the history of broadband multimedia communications and broadcasting. Among all, 5G technology comes with a better solution. Since it uses Non-Orthogonal Multiple Access (NOMA). NOMA uses the same resource blocks to transmit multiple

user's signals from the transmitter to the receiver. This will reduce delays, accommodate larger population satisfy traffic demands and reduces the time lag in communication. NOMA adopts super-position coding at the transmitter and successive interference cancellation at the receiver. During the transmission of signal in NOMA, at the transmitter side, it performs super position-coding technique and at the receiver side, it introduces some controllable interference like Successive Interference Cancellation (SIC), thus allows multiplexing users in the power domain.

CHAPTER-2

LITERATURE SURVEY

In[1] The non-orthogonal multiple access (NOMA) system is a feasible alternative for 5G technology because of its inherent benefit of high spectrum efficiency. This research provides closed-form BER formulas for near and remote users of the examined downlink NOMA. Monte Carlo simulations back up the study's conclusions.

In[2] This book covers existing and novel mutual authentication, random access, and waveform design approaches for 5G wireless networks and beyond. This book presents the most recent research from academia, business, and standards organizations. It's based on cutting-edge research papers, measurements, and experimental data gathered from a variety of sources.

In[3] For fifth-generation (5G) wireless networks, NOMA is a fundamental enabling technology. Based on the idea of serving and recognizing in the same block, as well as time frame, subcarrier, and spreading code. The most recent NOMA research and breakthroughs, as well as their applications, are included in this study.

In[4] For fifth-generation (5G) wireless networks, NOMA is a fundamental enabling technology. Based on the idea of serving and recognising in the same block, as well as time frame, subcarrier, and

spreading code. The most recent NOMA research and breakthroughs, as well as their applications, are included in this study.

In[5] With low-cost single-antenna terminals, multi-user MIMO may be accomplished. Across active terminals, a considerable number of service antennas are running in time-division duplex. By focusing radiation into progressively smaller regions, more antennas help to increase throughput and radiated energy efficiency. A propagation environment with asymptotically orthogonal channels is required for expected performance.

In[6] For 5G networks, non-orthogonal multiple access (NOMA) is expected to be a feasible multiple access strategy. First, the ergodic capacity maximisation issue is explored for Rayleigh fading multiple-input multiple-output (MIMO) NOMA systems using statistical channel state information at the transmitter

CHAPTER-3

METHODOLOGY

CODING TECHNIQUES

3.1 NOMA

Non-orthogonal multiple access (NOMA) principle emerges as a solution to improve the spectral efficiency while allowing some degree of multiple access interference at receivers. In NOMA, multiple users can utilize non-orthogonal resources concurrently by yielding a high spectral efficiency while allowing some degree of multiple access interference at receivers.

NOMA uses superposition coding at the transmitter, to combine the all user's signals into a single signal. Successive interference cancellation (SIC) receiver can separate the respective user's signal both in the uplink and in the downlink channels.

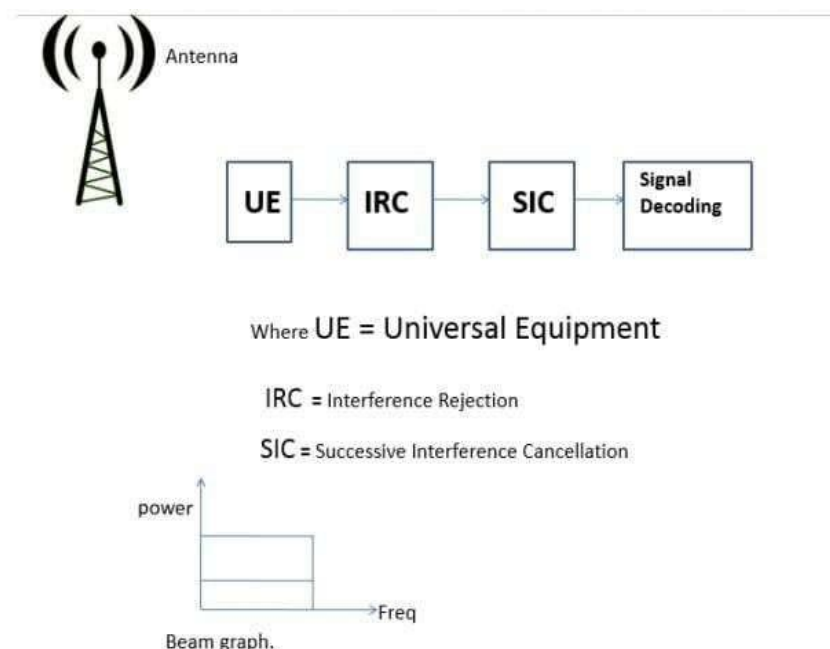


Fig 1: Block Diagram showing NOMA transmission in DL mode

Superposition coding at the transmitter and successive interference cancellation (SIC) at the receiver makes it possible to utilize the same spectrum for all users. At the transmitter site, all the individual information signals are superimposed into a single waveform, while at the receiver, SIC decodes the signals one by one until it finds the desired signal.

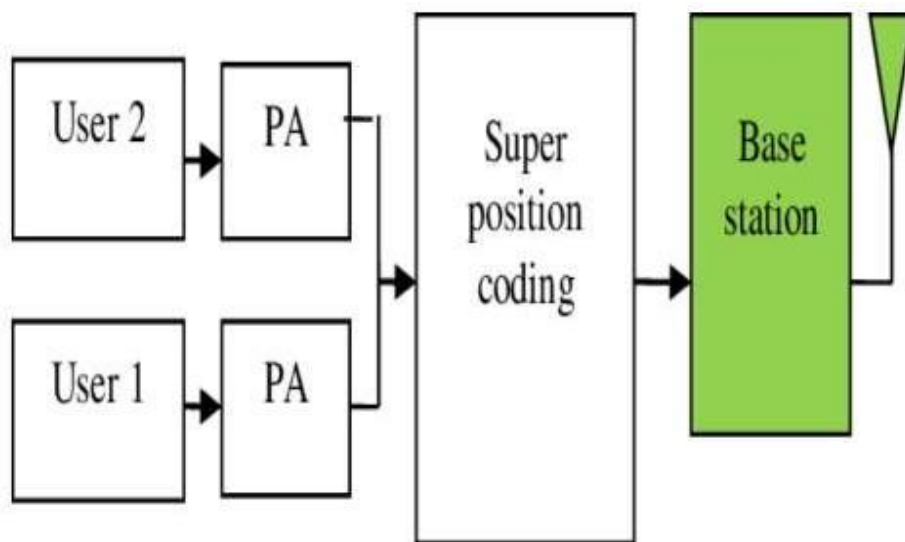
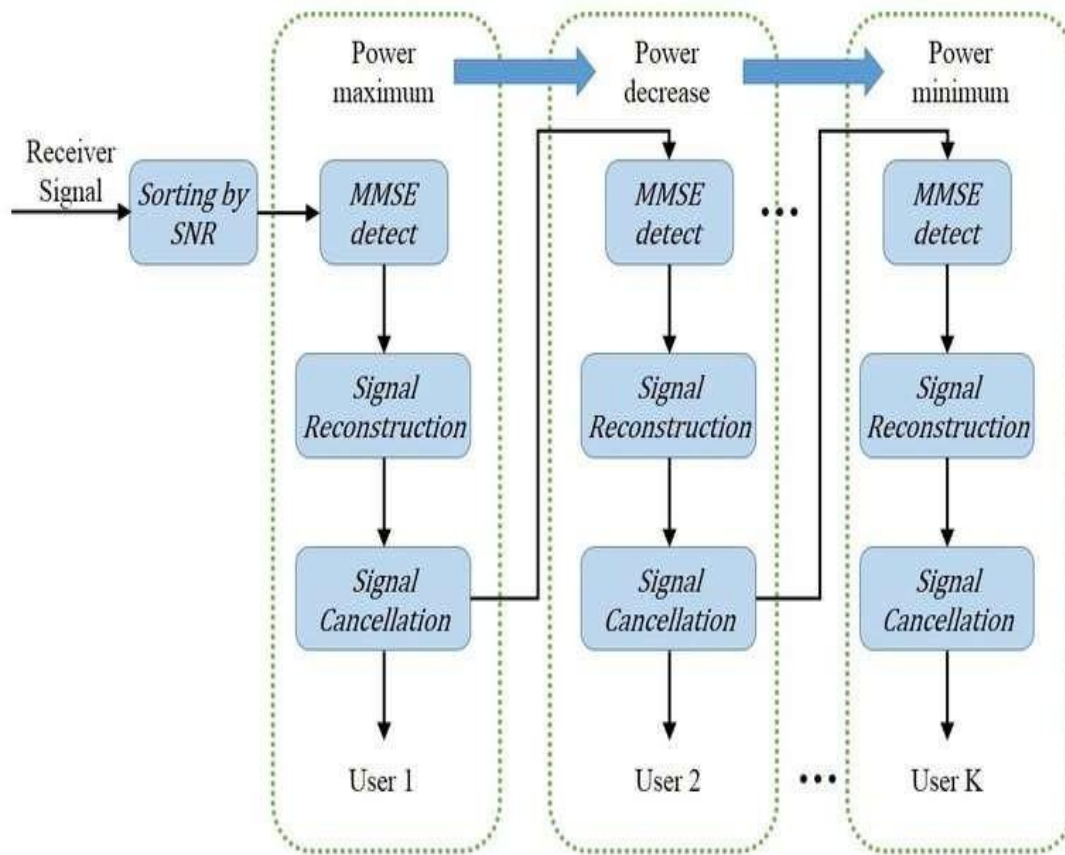


Fig. 1. Block diagram of NOMA in transmission section

At the transmitter side, each users signal is allocated with different power allocation coefficients and then it send the signals to the super position coding block to combine all the signals to a single signal and it is transmitted to base station.

At the receiver side, the signal is sorted according to SNR. Based on dominant power allocation coefficient among users, decoding takes place. The user with dominant power coefficient will decode first and then second dominant user decode next and so on till the last user.



In general, NOMA schemes can be classified into two types: power-domain multiplexing and code-domain multiplexing. In power-domain multiplexing, different users are allocated different power coefficients according to their channel conditions in order to achieve a high system performance. In particular, multiple user's information signals are superimposed at the transmitter side. At the receiver side successive interference cancellation (SIC) is applied for decoding the

signals one by one until the desired user's signal is obtained, providing a good trade-off between the throughput of the system and the user fairness.

In code-domain multiplexing, different users are allocated different codes and multiplexed over the same time-frequency resources.

In addition to power-domain multiplexing and code-domain multiplexing, there are other NOMA schemes such as pattern division multiple access (PDMA) and bit division multiplexing (BDM). Although code-domain multiplexing has a potential to enhance spectral efficiency, it requires a high transmission bandwidth. Considerable changes are not required on the existing networks, also, it does not require additional bandwidth in order to improve spectral efficiency.

In NOMA the same frequency resource is assigned to multiple mobile users, with good and bad channel conditions, at the same time. Hence, the resource assigned for the weak user is also used by the strong user, and the interference can be mitigated through SIC processes at user's receivers. Therefore, the probability of having improved spectral efficiency and a high throughput will be considerably increased.

NOMA can serve multiple users with different channel conditions simultaneously; therefore, it can provide improved user fairness, lower latency, and higher massive connectivity. And NOMA is also compatible with the current and future communication systems since it does not require significant modifications on the existing architecture.

3.2 MIMO-OMA

Different multiple access schemes are acquired from First Generation (1G) to Fourth Generation (4G). All the proposed schemes from 1G to 4G are common in one particular subject. That is the signals that are transmitted from the transmitter to the receiver must be orthogonal to each other. OMA uses orthogonal resource allocation so that only single user can use the resource block. When OMA is incorporated with MIMO system to get high diversity gain and less BER. In MIMO-OMA resource block is divided based on number of user's. Let us consider time division, then total time is equally divided into number of users and in first time slot, all the transmitter antennas transmits the first user signal and then in second time slot all the transmitter antennas transmits the second user signal to the second user and it continues till the last user respectively.

In MIMO-OMA the message passes from the transmitted to the receiver in different time slots, Like in a certain time slot a particular message will be travelled. Likewise the remaining message will be travelled. In terms of sum-rate the MIMO-OMA is low compared to the MIMO-NOMA for a simple two user scenario.

3.3 MIMO-NOMA

MIMO technologies have a significant capability of increasing capacity as well as improving error probability of wireless communication systems.

In order to achieve a balance between the maximum number of mobile users and the optimal achievable sum rate in MIMO-NOMA systems, sum rate has been represented through two ways. The first approach targets the optimization of power partition among the user clusters. Another approach is to group the users in different clusters such that each cluster can be allocated with orthogonal spectrum resources according to the selected user grouping algorithms.

The application of multiple-input multiple-output (MIMO) technologies to NOMA is important since the use of MIMO provides additional degrees of freedom for further performance improvement. MIMO can be used for either spatial multiplexing (increase achievable rate) or diversity gain (decrease BER).

As multiple techniques can achieve multi-stream beamforming coding diversity, the use of MIMO techniques brings flexible dimension for performance improvements. There are two techniques in MIMO-NOMA Zero forcing and block diagonalization based on zero-forcing (ZF) technique, low complexity linear precoding scheme can reduce inter-interference when channel transmission occurs. Block diagonalization (BD) is typical linear precoding technique with implemented in the downlink of multiuser MIMO systems.

The performance of downlink MIMO-NOMA network for a simple case of two users, that is, one cluster, is introduced. In this case, MIMO-NOMA provides a better performance than MIMO-OMA in terms of both the sum rate and ergodic sum rate. Also, it is shown that for a more practical case of multiple users, with two users allocated into a cluster and sharing the same transmit beamforming vector. MIMO-NOMA can achieve better outage performance than conventional MIMO-OMA, even for users that suffer strong co-channel interference.

3.4 SYSTEM MODEL OF NOMA

Consider a 2×1 downlink NOMA system. Let US1 and US2 be the user 1 and user 2 respectively. US1 be the far user or weak user and US2 be the near user or strong user. Let $D1$ and $D2$ denotes the distances of US1 and US2 respectively from transmitter. We assume, $D1 > D2$. Let $B1$ and $B2$ be the information intended for both users. Let us consider the Rayleigh fading channel between the transmitter antenna and receiver antenna. Following are the notations of NOMA, let C_n denote the fading channel between the transmitter antenna and near user and C_f denote the fading channel between the transmitter and the far user receiver antenna. At the transmitter side super position coding takes place to get the total signal transmitted, with different power allocation coefficients. Let TP be the total power for transmission of signals from transmitter to respective receiver antenna or user. The total signal transmitted by transmitter antennas for 2×1 NOMA is given by,

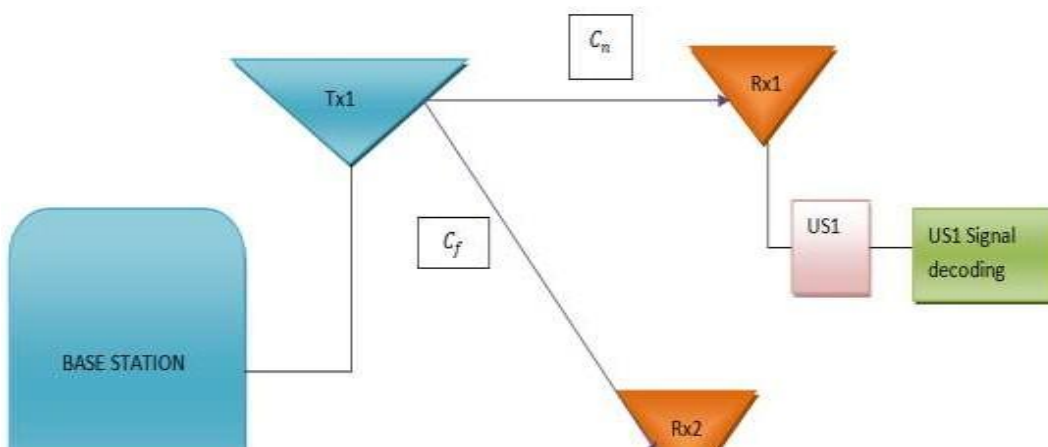


Fig 3.4 Showing NOMA Downlink system model with one BS and two distinct users.

In NOMA superposition coding is performed at the transmitter side and then the signal is transmitted to the receiver through the Rayleigh fading channel.

— — —

$$X_{\text{NOMA}} = \sqrt{P}(\sqrt{\alpha_1}x_1 + \sqrt{\alpha_2}x_2) \quad (1)$$

Then, at the receiver end, SIC is performed. After that, respective user information decoding takes place. In NOMA, the transmitter at the transmitter side transmits the total signal to all the receivers simultaneously. The signal received by the US1 receiver is given by,

$$Y_{1,\text{NOMA}} = C_{f1}X_{\text{NOMA}} + N_1 \quad (2)$$

Similarly, the signal received by US2 at the receiver side is given by,

$$Y_{2,\text{NOMA}} = C_{f2}X_{\text{NOMA}} + N_2 \quad (3)$$

Where C_f is the Rayleigh fading channel coefficient between transmitter and US1, C_n is the Rayleigh fading channel coefficient between transmitter and US2 receiver.

Decoding at User 1

Now, US1 has to decode B1 from the signal received by the US1 receiver. Since US1 is the weak signal, B1 is allocated with more power. Therefore, US1 can directly decode B1 from Y_{NOMA} , treating

B2 as interference. The SNR 1-NOMA equation is given by,

$$SNR_{1,NOM} = \frac{TP * p_1 |C_f|^2}{TP * p_1 |C_f|^2 + \sigma^2}$$

(4)

Therefore, US1 achievable rate is given by,

$$R_{1,NOM} = R_2(1 + R_{1,NOM}) \quad (5)$$

Decoding at User 2

User 2 must decode B2 from Y₂-NOMA. Since US2 is a strong user, B2 is allocated with less power. Therefore, in Y₂-NOMA, the power allocated to B1 is dominating. So, US2 will first perform direct decoding on Y₂-NOMA to obtain B1. Then SIC is carried out to remove B1. Then B2 is decoded. The SNR equation for US2 [1] is given by,

$$SNR_{2,NOM} = \frac{TP * p_2 |C_n|^2}{\sigma^2} \quad (6)$$

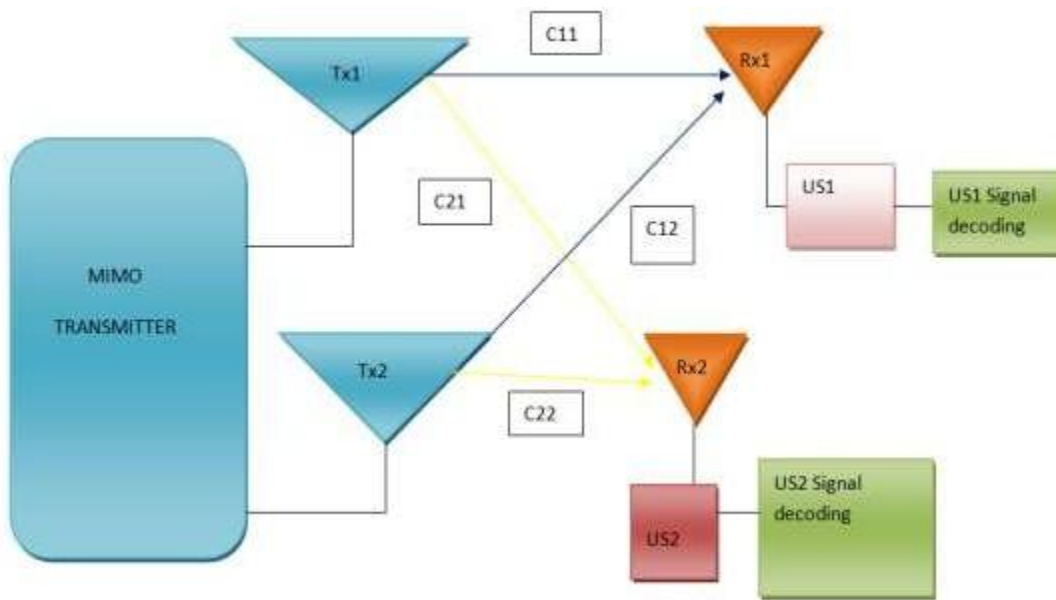
Therefore, the achievable rate for US2 [1] is given by,

$$R_{2,NOM} = \log_2(1 + SNR_{2,NOM}) \quad (7)$$

3.5 SYSTEM MODEL OF MIMO-OMA

Consider a 2 x 1 MIMO downlink system. Let us consider US1 and US2 to be user 1 and user 2 respectively. A far user or weak user is denoted by US1 and the near user or strong user is denoted by US2. Let D1 and D2 denote the distances of US1 and US2 respectively, from the MIMO transmitter. We assume, D1>D2. Let B1

and B_2 be the information intended for both users. The notations of the MIMO system are as follows, fading channel between t^{th} transmitter antenna and r^{th} receiver antenna is given by C_{rt} . Superposition coding takes place to get the total signal transmitted with different power allocation coefficients, at the transmitter side. The total PA coefficients must be equal to 1. Let us consider, TP to be the total power for the transmission of signals from the transmitter to the respective receiver antenna or user. Here in MIMO-OMA time is divided between the users. Both the transmitting antennas transmit the signal for user 1 in first time slot and then to the next user, thus time is divided among the users.



C_{rt} =Channel from t^{th} transmit antenna to r^{th} receiver antenna

Fig3.5 Showing MIMO-OMA Downlink system model having one BS and two distinct users

To know how MIMO-NOMA performs well, we are going to

compare MIMO-NOMA with MIMO-OMA and NOMA. In MIMO-OMA, suppose we consider 2 users then the transmission time is divided into two slots. In the first time slot, both transmitter antennas transmit to the US1 signal and in the second time slot, both the transmitter antennas transmit the US2 signal. The signal transmitted by both the transmitter antennas at the first time slot is $TP*B1$. The signal received by US1 in the first time slot is given by,

—

$$y_{1,OMA} = \sqrt{P} \mathbf{h}_1^H (\mathbf{h}_{11} + \mathbf{h}_{12}) x_1 + n_1 \quad (8)$$

Similarly, the signal transmitted by both the antennas in the second time slot to US2 is $TP*B2$ and the signal received by US2 is given by,

—

$$y_{2,OMA} = \sqrt{P} \mathbf{h}_2^H (\mathbf{h}_{21} + \mathbf{h}_{22}) x_2 + n_2 \quad (9)$$

The SNR at US1 and US2 are given by,

$$SNR_{1,OMA} = \frac{TP |\mathbf{h}_{11} + \mathbf{h}_{12}|^2}{\sigma^2} \quad (10)$$

$$SNR_{2,OMA} = \frac{TP |\mathbf{h}_{21} + \mathbf{h}_{22}|^2}{\sigma^2} \quad (11)$$

1;2 achievable rates for MIMO-OMA are given by,

□□

$$= \frac{1}{2} \log_2 (1 + SNR_{1,OMA})$$

□□

$$= \frac{1}{2} \log_2 (1 + SNR_{2,OMA})$$

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—

$$r_{1,OMA} = \sqrt{P}(\alpha_{11} + \alpha_{12}) + n_1 \quad (8)$$

Similarly, the signal transmitted by both the antennas in the second time slot to US2 is $TP*B2$ and the signal received by US2 is given by,

—

$$r_{2,OMA} = \sqrt{P}(\alpha_{21} + \alpha_{22}) + n_2 \quad (9)$$

The SNR at US1 and US2 are given by,

$$\frac{r_{1,OMA}^2}{\sigma^2} = \frac{P(\alpha_{11} + \alpha_{12})^2}{\sigma^2} \quad (10)$$

$$\frac{r_{2,OMA}^2}{\sigma^2} = \frac{P(\alpha_{21} + \alpha_{22})^2}{\sigma^2} \quad (11)$$

1;2 achievable rates for MIMO-OMA are given by,

□□

$$= \frac{1}{2} \log_2 (1 + \frac{P_1}{P_2} \frac{1}{1 + \frac{P_1}{P_2}})$$

□□

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3.6 SYSTEM MODEL OF MIMO-NOMA

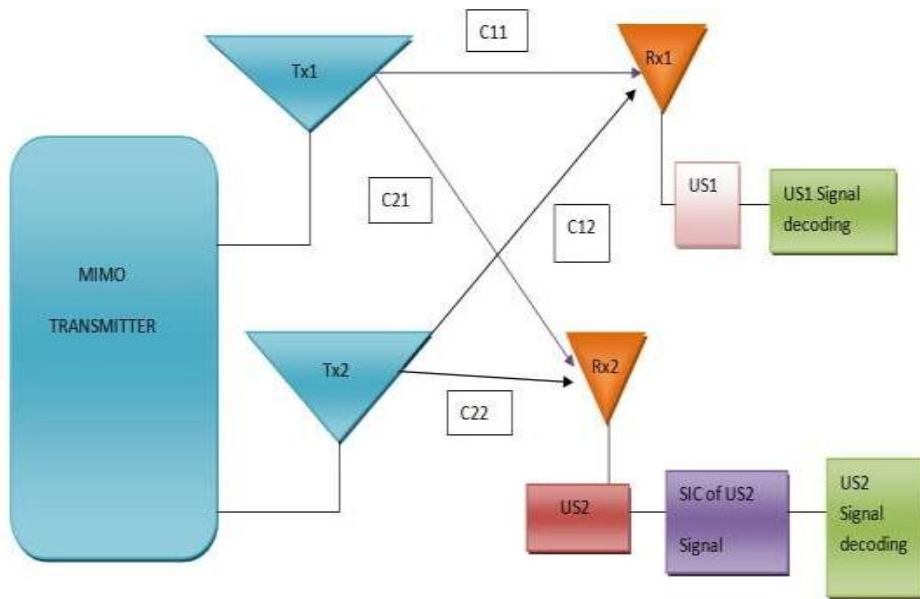
Consider a 2 x 1 MIMO downlink system. Let us consider US1 and US2 to be user 1 and user 2 respectively [2]. A far user or weak user is denoted by US1 and the near user or strong user is denoted by US2. Let D1 and D2 denote the distances of US1 and US2 respectively, from the MIMO transmitter. We assume, D1>D2. Let B1 and B2 be the information intended for both users. The notations of the MIMO system are as follows, fading channel between tth transmitter antenna and rth receiver antenna is given by C_{rt}.

Superposition coding takes place to get the total signal transmitted with different power allocation coefficients, at the transmitter side. The total PA coefficients must be equal to 1. Let us consider, TP to be the total power for the transmission of signals from the transmitter to the respective receiver antenna or user. The total signal transmitted by transmitter antennas for 2 x1 MIMO systems is given by

— — —

$$s = \sqrt{\alpha} s_1 + \sqrt{1-\alpha} s_2 \quad (14)$$

Where p_1 and p_2 are the NOMA power allocation coefficients. $p_1 >$



p_2 , because US1 is a distant user or weak user.

Fig 3.6 Showing MIMO Downlink system model having one Base Station and two distinct users.

The total signal is transmitted to each receiver in multiple paths by each transmitter. At the receiver end, the same signal is received by both the antennas. Therefore, the received signal at US1 is given by,

$$r_{1, \text{received}} = h_{11}x_1 + h_{12}x_2 + n_1 = (h_{11} + h_{12})x_1 + n_1 \quad (15) \text{ Similarly}$$

$$r_{2, \text{received}} = h_{21}x_1 + h_{22}x_2 + n_2 = (h_{21} + h_{22})x_1 + n_2 \quad (16) \text{ Where N}$$

Decoding at User1

At US1, the intended data for the US1 signal has the dominating PA coefficient, so direct decoding of the US1 signal takes

place. The information intended for US2 is treated as interference at the US1 receiver. The SNR of a signal is given by the ratio of the original signal intended for the respective user, to the interference. Then SNR1 for US1 in decoding B1 is given by,

$$SNR_{1,MIMO} = \frac{(TP * p_1 |C_{11} + C_{12}|^2)}{(TP * p_2 |C_{11} + C_{12}|^2) + \sigma^2} \quad (17)$$

Therefore, the achievable rate for US1 is given by, $AR_{1,MIMO} = \log_2(1 + SNR_{1,MIMO})$

$$AR_{1,MIMO} = \log_2(1 + SNR_{1,MIMO}) \quad (18)$$

Decoding at User2

Whereas at US2, US1 signal information is allocated with more power, so decoding of US1 signal takes place directly by treating US2 signal as interference. The SNR at the US1 for direct decoding of B1 is given by,

$$\begin{aligned} SNR_{1,MIMO} &= \frac{(TP * p_1 |C_{11} + C_{12}|^2)}{(TP * p_2 |C_{11} + C_{12}|^2) + \sigma^2} \\ &= \frac{(TP * p_1 |C_{11} + C_{12}|^2)}{(TP * p_2 |C_{11} + C_{12}|^2) + \sigma^2} \end{aligned} \quad (19)$$

After performing SIC (successive interference cancellation), the US1 data is removed from Y, and then the remaining signal is

decoded to get the US2 signal. The SNR equation for US2 to decode its signal is given by,

$$\frac{SNR_{2,MIM}}{Q} = \frac{TP + p_2 |G_{21} + G_{22}|^2}{\sigma^2} \quad (20)$$

The achievable rates at US2 for decoding B1 and B2 are given by,

$$\begin{aligned} AR_{12,MIMO} &= \log_2(1 + SNR_{12,MIMO}) \\ AR_{2,MIMO} &= \log_2(1 + SNR_{2,MIMO}) \end{aligned} \quad (21)$$

PERFORMANCE ANALYSIS

3.7 POWER ALLOCATION

Wireless data networks are increasingly becoming an important part of next generation network infrastructure. The objective of these networks is to provide with anytime, anywhere data access. The computing and storage capabilities of these devices cover a wide spectrum. Power consumption has become an important concern when it comes to the implementation phase of wireless devices. Therefore, power management is one of the challenging problems in wireless communications.

In wireless networks, power allocation is an effective technique for prolonging the lifetime of network terminals. Generally optimum power allocation improves the efficiency of wireless systems. When power allocation is properly done, source information can reach the destination

efficiently. The problem of power allocation in relay assisted wireless system is investigated here with Maximal Ratio Combining (MRC) at the destination terminal. Based on the SNR and BER, power is allocated to the network terminals using water filling power allocation scheme. It is observed that water filling power allocation scheme allocates more power for the network terminals with less noise. Also based on the users channel capacity optimized power allocation is done. Hence the channel serves good enough in transmitting the source information to the destination.

The amount of power that is allocated to the user data to transmit the respective signal from the transmitter to the receiver is called power allocation. By implementing the optimum power allocation process, the efficiency of the wireless communications system has been improved. To transmit the signal from the transmitter antenna to the receiver antenna, power allocation coefficients are required, as NOMA is based on a power domain technique. The user which is distant from the transmitter antenna is allocated with high power and the user which is close to the transmitter is allocated with less power coefficient. The summation of all power allocation coefficients must be 1.

3.8 CHANNEL CAPACITY

The maximum rate at which data can be transmitted over a communication path or channel is called channel capacity. The channel capacity depends upon four factors they are data rate (bps), bandwidth (Hz), noise (dB), error rate (in %).

Wireless channel is prone to multipath propagation and fading. Several channel models are available to capture the effects of fading. Each model deals with a particular scenario. One such model is Rayleigh fading model. Rayleigh fading model can be used when there is no line of sight (LOS) path between the transmitter and the receiver. In other words, all multipath components have undergone small scale fading effects like reflection, scattering, diffraction, shadowing etc. We are going to consider an extreme case of Rayleigh fading where each transmitted bit undergoes a different attenuation and phase shift due to multipath transmission. In other words, the channel changes for every bit.

To calculate channel capacity of each technique first distances for far user and near user should be taken and power allocation coefficients are allocated to each user. Far user must be allocated with more power than near user.

Channel capacity = $AR_1 + AR_2$ (23). Where
AR1 and AR2 are the achievable rates for user 1 and user 2.

In a mobile communication environment the channel is not time invariant and is slowly varying. This characteristic feature of the channel leads to a phenomenon called Fading. Fading channels induce rapid amplitude fluctuations in the received signal. If they are not compensated for then this will lead to serious performance degradation.

3.9 OUTAGE PROBABILITY

Outage probability is the ratio of the number of samples of the signal below the threshold to the total number of samples. If we increase the threshold power, the outage probability is increased, that is the probability of better detection of signal is decreased. In Information theory, outage probability of a communication channel is the probability that a given information rate is not supported, because of variable channel capacity. Outage probability is defined as the probability that information rate is less than the required threshold information rate. It is the probability that an outage will occur within a specified time period. Mathematically, the outage probability can be represented as the CDF of the SNR and is given by,

$$P_{out}(R) = \Pr [AR < R] \quad (24)$$

Where AR is considered as the achievable rate for the respective user and R is the target rate or the threshold value. If R is high blocking rate is more. To get OP below a maximum value, the SNR threshold value should be kept low. If AR is less than the target rate, then the outage occurs. In case the target rate is high then the outage probability reaches the maximum value as noise dominates the original information. To calculate the outage probability for a user, set the target rate and count the number of times that the calculated achievable rates dropped below the target rate and plot them. In MIMO-NOMA outage probability decreases as directivity gain increases, because in MIMO-NOMA multiple antennas are used to increase directivity gain. As it is evident, that the outage probability should be less to get a significant result.

4.0 Bit ErrorRate

The number of bit errors is the number of received bits of a data stream over a communication channel that have been altered due to noise, interference, distortion or bit synchronization errors. The bit error rate (BER) is the number of bit errors per unit time. The bit error ratio (also BER) is the number of bit errors divided by the total number of transferred bits during a studied time interval. Bit error ratio is a unit less performance measure, often expressed as a percentage. The bit error probability p_e is the expectation value of the bit error ratio. The bit error ratio can be considered as an approximate estimate of the bit error probability. This estimate is accurate for a long time interval and a high number of bit errors. The quality of digital wireless transmission is determined by the measurement of BER. If the BER of a system is low then, it provides better performance. Moreover, the different factors like channel noise, interference, distortion, and attenuation affect the BER during the transmission of a signal. Here, in this paper, we calculated BER by comparing the copy of the transmitted signal with the signal that is decoded at the receiver to the total number of bits, and then it is plotted against transmitted power.

Bit error rate is a key parameter that is used in assessing systems that transmit digital data from one location to another. BER is applicable to radio data links, Ethernet, as well as fiber optic data systems. When data is transmitted over a data link, there is a possibility of errors being introduced into the system. If this is so, the integrity of the system may be compromised. As a result, it is necessary to assess the performance of the system, and BER

provides an ideal way in which this can be achieved. BER assesses the full end to end performance of a system including the transmitter, receiver and the medium between the two.

CHAPTER-5

MATLAB

MATLABINTRODUCTION

MATLAB is a high performance language for technical computing. It integrates computation visualization and programming in an easy to use environment. MATLAB stands for matrix laboratory. It was written originally to provide easy access to matrix software developed by LINPACK (linear system package) and EISPACK (Eigen system package) projects. MATLAB is therefore built on a foundation of sophisticated matrix software in which the basic element is matrix that does not require pre dimensioning. Typical uses of MATLAB

1. Math and computation
2. Algorithm development
3. Data acquisition
4. Data analysis, exploration and visualization

The main features of MATLAB

1. Advanced algorithm for high performance numerical computation, especially in the Field matrix algebra
2. A large collection of predefined mathematical functions and the ability to define one's own functions.
3. Two-and three dimensional graphics for plotting and displaying data
4. A complete online help system
5. Powerful matrix or vector oriented high level programming language for individual applications.
6. Toolboxes available for solving advanced problems in several application areas.

MATLAB INSTALLATION

Installation Procedure for MATLAB Software

Redundant clients-Windows XP-

SP2/Vista/Windows 7/Windows 8

Download:

- win32.rar or win64.rar (for 64 bitsystem)
- PLP
- "license.dat" **from** <http://mirror.iitd.ernet.in/matlab> into Your DesiredFolder.

STEPS: Installation of MATLAB Software for window XP/VISTA / Windows 7/8

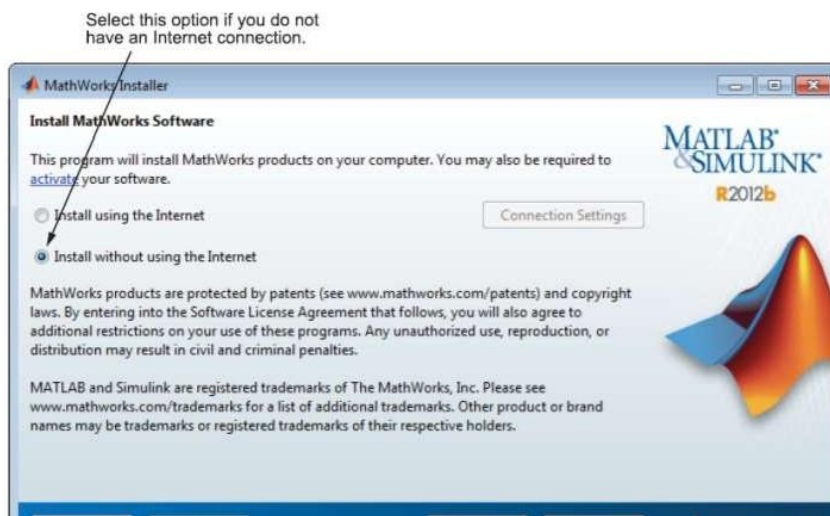
Get administrator privileges for the system on which you plan to install MATLAB. Use WinRAR to extract RAR file

Step1: Start the installer

For Windows, double-clicking on setup.exe

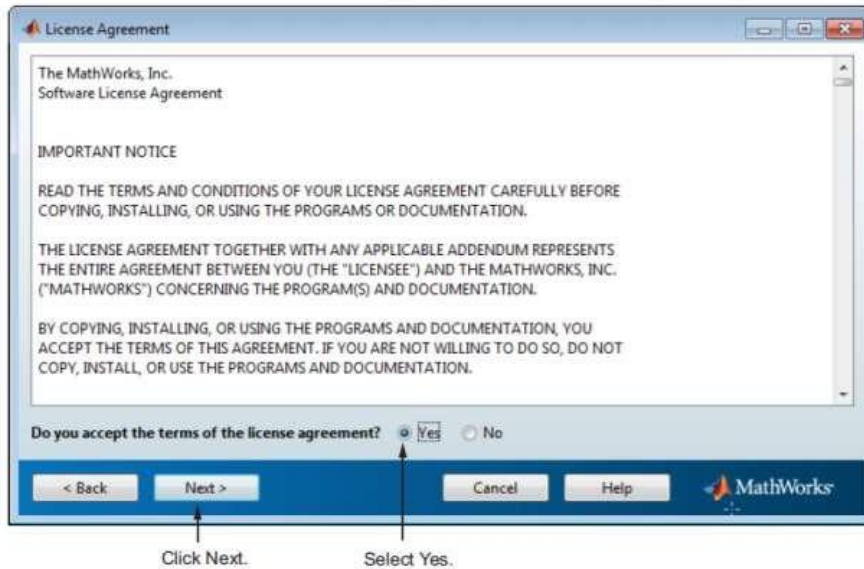
Step 2: Choose to Install Without Using the Internet

When it starts, the installer displays the following dialog box. Select the Install without using the Internet option and Click OK to proceed with installation.



Step 3: Review the License Agreement

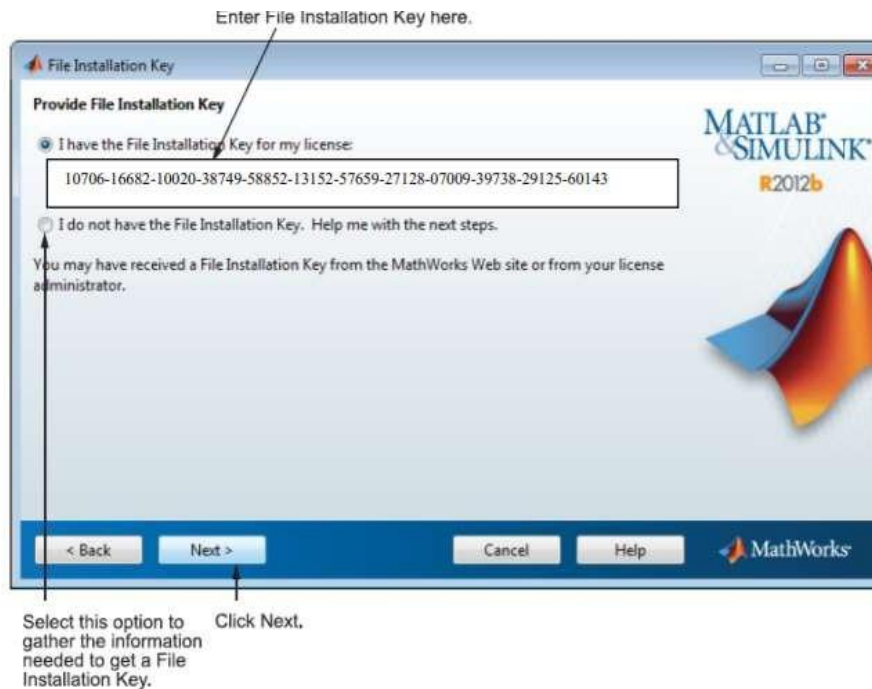
Review the software licensing agreement and, if you agree to its terms, click Yes.



Step 4: Enter the File Installation Key

Enter your File Installation Key and Click OK.

10706-16682-10020-38749-58852-13152-57659-27128-07009-39738-29125-60143



Step 5: Choose the Installation Type

In the Installation Type dialog box, specify whether you want to perform a Custom installation and click Next.



Step 6: Specify the Installation Folder

Specify the name of the folder where you want to install

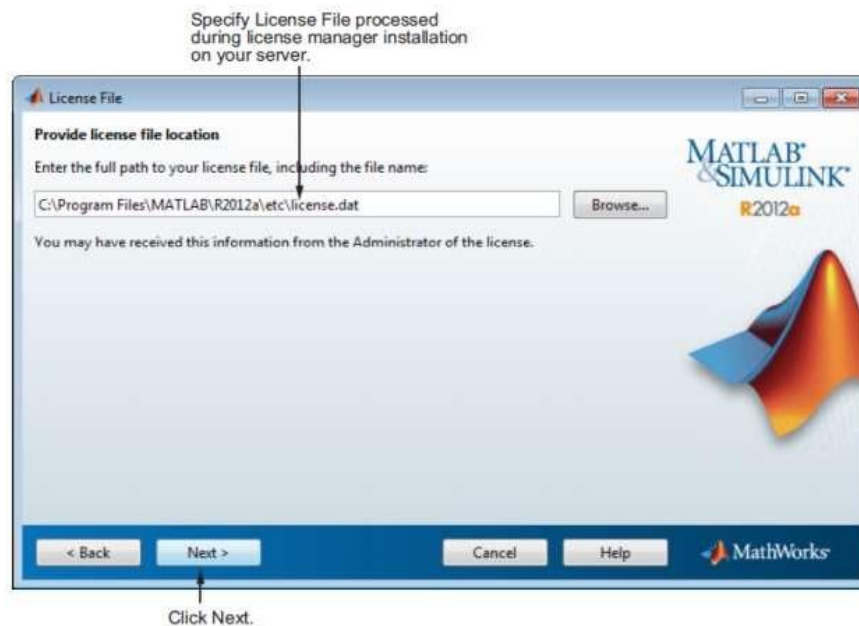
MathWorks products. Accept the default installation folder or click Browse to select a different one. If the folder doesn't exist, the installer creates it.

Step 7: Specify Products to Install (Custom Only)

Leave it by default and continue.

Step 8: Specify the Location of the License File

Enter the full path of your License File in the text box (or drag and drop the file) and click Next.

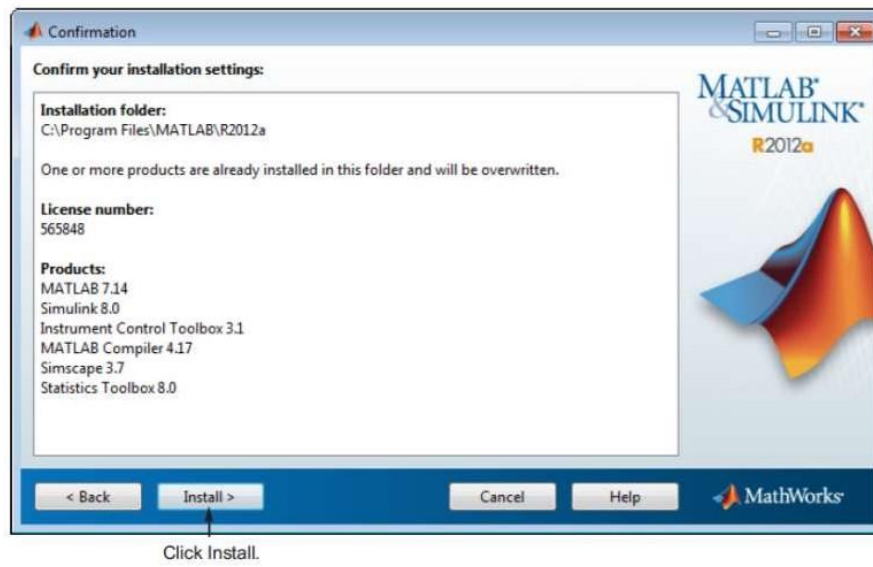


Step 9: Specify Installation Options (Custom Only)

After selecting installation options, click Next to proceed with the installation.

Step 10: Confirm Your Choices and Begin Copying

Files Before it begins copying files to your hard disk, the installer displays a summary of your installation choices. To change a setting, click Back. To proceed with the installation, click Install. As it copies files to your hard drive, the installer displays a status dialog box to show the progress of the installation.



Step 11: Complete the Installation

Step 11: Complete the Installation



The MATLABSYSTEM

The MATLAB System consists of five mainparts

- DevelopmentEnvironment:

This is the set of tools and facilities that help you use MATLAB functions and files. Many of these tools are graphical user interfaces. It includes the MATLAB desktop and Command Window, command history an editor and debugger, and browsers for viewing help the workspace, files, and the search path.

- The MATLAB Mathematical FunctionLibrary:

This is a vast collection of computational algorithms ranging from elementary functions, like sum sine, cosine, and complex arithmetic, to more sophisticated functions like matrix inverse, matrix Eigen values, Bessel functions, and fast Fourier transforms.

- The MATLABLanguage:

This is a high-level matrix/array language with control flow statements, functions, data structures, input/output, and object-oriented programming features. It allows both programming in the small to rapidly create quickly programs, and "programming in the large" to create large and complex application programs.

- Graphics: MATLAB has extensive facilities for displaying vectors and matrices as graphs, as well as annotating and printing these graphs. It includes high-level functions for two-dimensional and three dimensional data visualization, video processing,

animation, and presentation graphics. It also includes low-level functions that allow you to fully customize the appearance of graphics as well as to build complete graphical user interfaces on your MATLAB applications

- The MATLAB Application Program Interface(API):

This is a library that allows you to write C and Fortran programs that interact with MATLAB. It includes facilities for calling routines from MATLAB (dynamic linking), calling MATLAB as a computational engine, and for reading and writing MAT-files .

CHAPTER-4

RESULTS AND DISCUSSION

This project focuses on how MIMO-NOMA outperforms remaining techniques like NOMA and OMA w.r.t each parameter. The Simulation results shows a comparison of each parameter w.r.t the three different techniques. For the analysis of Power allocation, Channel capacity, Outage probability, and Bit error rate for three different techniques, MATLAB 2015a is used with the following specifications.

4.1 Power allocation

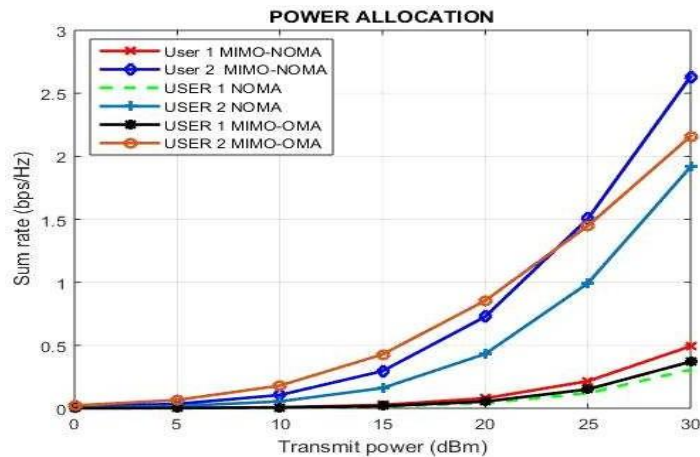


Figure:Power Allocation of MIMO-NOMA, NOMA, and OMA for two different users

Figure illustrates the power allocation of three different techniques - OMA, NOMA, and MIMO-NOMA. The PA of two users-far user and a near user is depicted in the figure. It is evident from the figure that the power allocation of three different techniques-OMA, NOMA, and MIMO-NOMA for two different users are observed and analysed close to each other. For near user which is allocated with less power gives better performance than the far user which is allocated with high power. Among all, MIMO-NOMA outperforms remaining techniques at larger values of transmit power. For both near and far users MIMO-NOMA beat the remaining techniques at transmit power.

4.2 Channel Capacity

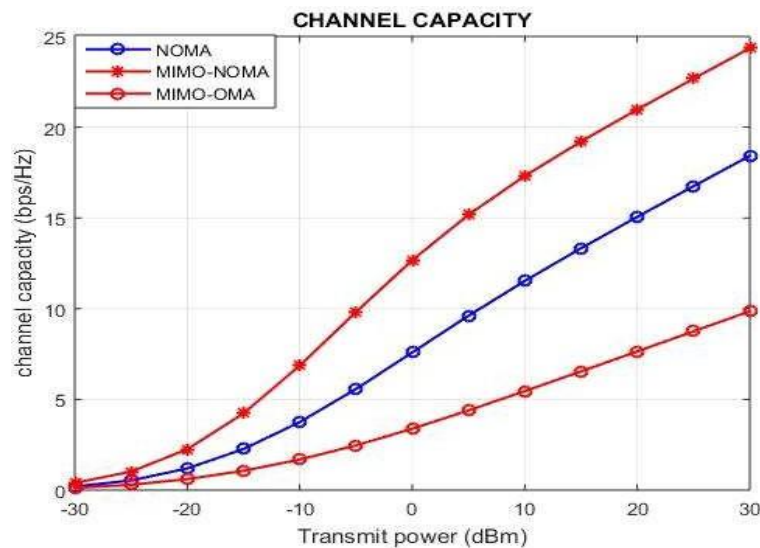


Figure:Channel Capacity of MIMO-NOMA, NOMA, and OMA for two different users

Figure depicts the effect of channel capacity for three different

techniques OMA, NOMA, and MIMO-NOMA. Among all MIMO-NOMA outperforms remaining techniques at all the values of transmit power at transmitting power.

4.3 Outage Probability

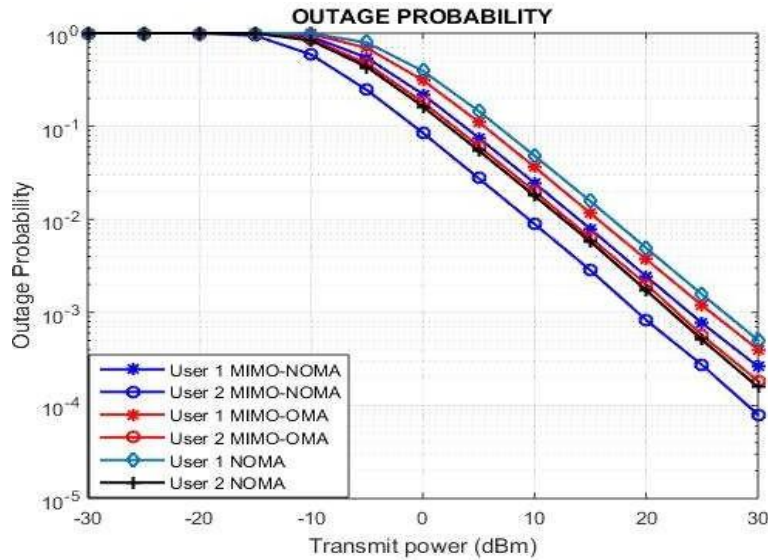


Figure: Outage Probability of MIMO-NOMA, NOMA, and OMA for two different users.

Figure shows the effect of outage probability for three different techniques-OMA, NOMA, and MIMO-NOMA. Among all, MIMO-NOMA outperforms remaining techniques at all the values of transmitting power it is clear that shown the Outage probability of MIMO-NOMA is less comparatively. Lesser the outage probability better would be the performance of the system.

1. Bit Error Rate

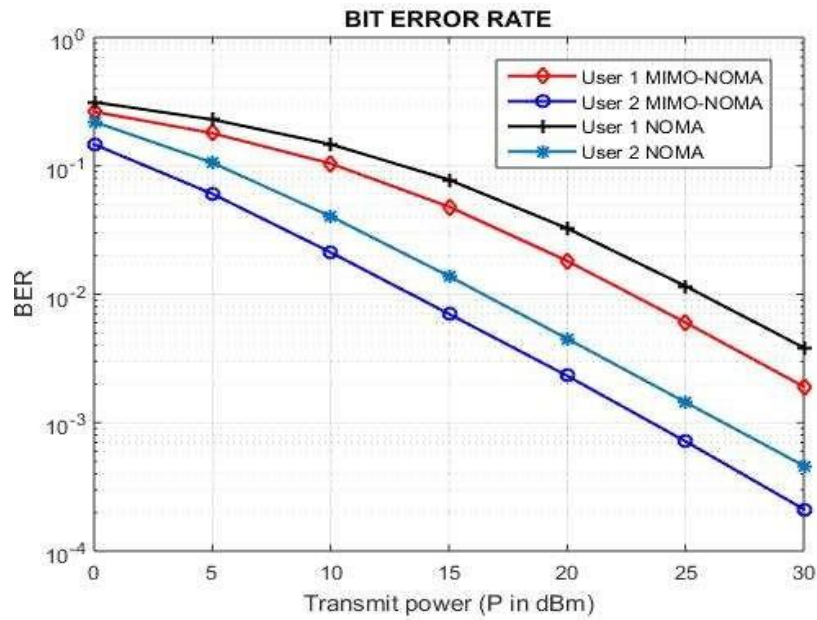


Figure: Bit Error Rate of MIMO-NOMA and NOMA for two different users.

Figure depicts the effect of Bit error rate for three different techniques-OMA, NOMA, and MIMO-NOMA. Among all MIMO-NOMA outperforms remaining techniques at all the values of transmit power, it is clear that the BER of MIMO-NOMA is less comparatively. Lesser the bit error rate better would be the performance.

CHAPTER-5

CONCLUSION AND FUTURE WORK

In OMA, Orthogonal multiple access techniques are used, because of that it can strict to limited users and it cannot satisfy the number of subscribers day by day, and also it cannot satisfy the low latency, massive connectivity and spectral efficiency. To satisfy the number of increasing subscribers NOMA technique is introduced. The performance of NOMA is better than OMA in latency, spectral efficiency, and massive connectivity because it uses the same resource blocks for different user's data transmission. But Bit error rate and SNR are not up to the mark. A MIMO system is incorporated with the NOMA technique to achieve less BER and good SNR. An individual sum-rate and SNR equations for each user are calculated and compared among MIMO-NOMA, NOMA, and OMA have been investigated and plotted. From that, we conclude that MIMO-NOMA achieves a higher individual rate than NOMA and OMA for a simple two-user scenario. For all four parameters-Power allocation, Channel capacity, Outage probability, and Bit error rate, a comparison has been done among all the three techniques in which MIMO-NOMA provides better performance. In this paper, we provide a complete analysis of how each system performs and are compared each system model with others

for a simple two-user scenario. For all four parameters, the Simulation results and performance of each technique are analysed.

Future Scope

In this paper we have used Rayleigh fading channel between the transmitter and the receiver. In future this work can be extended by replacing Rayleigh channel with Rician and AWGN channels and predicting the channel which provides better performance.

• References

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2. Hongji Huang, Song Guo, Guan Gui, Zhen Yang, Jianhua Zhang, Hikmet Sari, Fumiyuki Adachi "Deep Learning for Physical-Layer 5G Wireless Techniques: Opportunities, Challenges, Solutions" IEEE, 2020.
3. Hyoungju Ji, Sunho Park, Jeongho Yeo, Younsun Kim, Juho Lee, Byonghyo Shim "Ultra-Reliable and Low- Latency Communications in 5G Downlink: Physical Layer Aspects" IEEE, 2018.

4. Qi Sun, Shuangfeng Han, Chin-Lin I, and Zhengang Pan, "On the Ergodic Capacity of MIMO NOMA Systems," IEEE WIRELESS COMMUNICATION, VOL. X, NO. X,XX 2015.
5. E.G. Larsson, O.Edfors, F. Tufvesson, and T. L. Marzetta, "Massive MIMO for next-generation wireless systems," IEEE Communications Magazine, vol. 52, no. 2, pp. 186–195, 2014.

- **Appendices**

A. Sample Codes

1.Power Allocation

```
%% RELEIGH FADING CHANNEL POWER ALLOCATION  
clc; clear variables; close all;
```

```
%Distances  
d1 = 5000; d2 = 2000;
```

```
% power allocation coefficients  
b1= 0.75;  
b2= 0.25;
```

```

N = 5*10^5;

eta = 4;%Path loss exponent

%Rayleigh fading channels
h11 = sqrt(d1^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h12 = sqrt(d1^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h21 = sqrt(d2^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h22 = sqrt(d2^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);

h1 = h11+h12;
h2 = h21+h22;

hf = sqrt(d1^-eta)*(randn(1,N) + 1i*randn(1,N))/sqrt(2);
hn = sqrt(d2^-eta)*(randn(1,N) + 1i*randn(1,N))/sqrt(2);

%Channel gains
g1 = (abs(h1)).^2;
g2 = (abs(h2)).^2;

gf = (abs(hf)).^2;
gn = (abs(hn)).^2;
%Transmit power
Pt = 0:5:30; %in dBm
pt = (10^-3)*db2pow(Pt); %linear scale

BW = 10^6; %bandwidth
%Noise power
No = -174 + 10*log10(BW); %in dBm
no = (10^-3)*db2pow(No); %in linear scale

%Target rates
R1 = 1;
R2 = 3;
epsilon = ((2^R1)-1) ;
for u = 1:length(pt)
    a1 = epsilon*(no + pt(u)*g1)./(pt(u)*g1*(1+epsilon));
    a1(a1>1) = 0;
    a2 = 1 - a1;
    %Achievable rates for MIMO-NOMA Fair PA
    R1n = log2(1 + pt(u)*a1.*g1./(pt(u)*a2.*g1 + no));
    R12n = log2(1 + pt(u)*a1.*g2./(pt(u)*a2.*g2 + no));
    R2n = log2(1 + pt(u)*a2.*g2/no);

```

```

R1o = 0.5*log2(1 + pt(u)*g1/no);
R2o = 0.5*log2(1 + pt(u)*g2/no);

%Achievable rates for MIMO-NOMA Fixed PA
R1nf = log2(1 + pt(u)*b1.*g1./(pt(u)*b2.*g1 + no));
R12nf = log2(1 + pt(u)*b1.*g2./(pt(u)*b2.*g2 + no));
R2nf = log2(1 + pt(u)*b2.*g2/no);

%Sum rates
Rn(u) = mean(R1n+R2n); %MIMO-NOMA
Ro(u) = mean(R1o+R2o);
R1o_av(u) = mean(R1o); %MIMO-OMA USER 1 (WEAK)
R2o_av(u) = mean(R2o); %MIMO-OMA USER 1 (WEAK)
Rnf(u) = mean(R1nf+R2nf);
R1(u)=mean(R1nf);
R2(u)=mean(R2nf);

end
for u = 1:length(pt)
    a1 = epsilon*(no + pt(u)*gf)./(pt(u)*gf*(1+epsilon));
    a1(a1>1) = 0;
    a2 = 1 - a1;
    %Achievable rates for NOMA
    C1 = log2(1 + pt(u)*a1.*gf./(pt(u)*a2.*gf + no));
    C2 = log2(1 + pt(u)*a2.*gn/no);

    C_sum(u) = mean(C1+C2);
    C1(u) = mean(C1);
    C2(u) = mean(C2);
    %Sum rate of fixed PA NOMA
    C1f = log2(1 + pt(u)*b1.*gf./(pt(u)*b2.*gf + no));
    C2f = log2(1 + pt(u)*b2.*gn/no);
    C_sumf(u) = mean(C1f+C2f);
    C11(u) = mean(C1f);
    C22(u) = mean(C2f);
end
figure();
plot(Pt, R1,'-r','linewidth',2); hold on; grid on;
xlabel('Transmit power (dBm)');
ylabel('Sum rate (bps/Hz)');
plot(Pt, R2,'-b','linewidth',2); hold on; grid on;
plot(Pt, C11,'-g','linewidth',2); hold on; grid on;
plot(Pt,C22,'-', 'linewidth',2); hold on; grid on;
plot(Pt,R1o_av,'-k','linewidth',2); hold on; grid on;
plot(Pt,R2o_av,'-', 'linewidth',2); hold on; grid on;

```

```

title('POWER ALLOCATION');
legend('User 1 MIMO-NOMA','User 2 MIMO-NOMA','USER 1 NOMA','USER 2 NOMA','USER 1 MIMO-OMA','USER 2 MIMO-OMA');

```

2. Channel Capacity

```

clc; clear variables; close all;

```

```

%Distances
d1 = 500; d2 = 200;

```

```

%Power allocation coefficients
a1 = 0.75; a2 = 0.25;

```

```

N = 5*10^5;
eta = 4;%Path loss exponent

```

```

%Rayleigh fading channels
h11 = sqrt(d1^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h12 = sqrt(d1^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h21 = sqrt(d2^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h22 = sqrt(d2^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);

```

```

h1 = h11+h12;
h2 = h21+h22;

```

```

%Channel gains
g1 = (abs(h1)).^2;
g2 = (abs(h2)).^2;

```

```

hf = sqrt(d1^-eta)*(randn(1,N) + 1i*randn(1,N))/sqrt(2);
hn = sqrt(d2^-eta)*(randn(1,N) + 1i*randn(1,N))/sqrt(2);

```

```

gf = (abs(hf)).^2;
gn = (abs(hn)).^2;

```

```

%Transmit power
Pt = -30:5:30; %in dBm
pt = (10^-3)*db2pow(Pt); %linear scale

```

```

BW = 10^6; %bandwidth
%Noise power
No = -174 + 10*log10(BW); %in dBm
no = (10^-3)*db2pow(No); %in linear scale

```

```

%Target rates
R1 = 1;
R2 = 3;

```

```

p1n = zeros(1,length(pt));
p2n = zeros(1,length(pt));
p1o = zeros(1,length(pt));
p2o = zeros(1,length(pt));

for u = 1:length(pt)

    %Achievable rates for MIMO-NOMA
    R1n = log2(1 + pt(u)*a1.*g1./(pt(u)*a2.*g1 + no));
    R12n = log2(1 + pt(u)*a1.*g2./(pt(u)*a2.*g2 + no));
    R2n = log2(1 + pt(u)*a2.*g2/no);

    %NOMA
    R11n = log2(1 + pt(u)*a1.*gf./(pt(u)*a2.*gf + no));
    R121n = log2(1 + pt(u)*a1.*gn./(pt(u)*a2.*gn + no));
    R21n = log2(1 + pt(u)*a2.*gn/no);

    %Achievable rates for MIMO-OMA
    R1o = 0.5*log2(1 + pt(u)*g1/no);
    R2o = 0.5*log2(1 + pt(u)*g2/no);

    %%Achievable rates for NOMA
    %NOMA capacity calculation
    C_noma_1 = log2(1 + pt(u)*a1.*gf./(pt(u)*a2.*gf + no)); %User 1
    C_noma_2 = log2(1 + pt(u)*a1.*gn./(pt(u)*a2.*gn+no));
    C_noma_3 = log2(1 + pt(u)*a2.*gn/no); %User 2
    C_noma_sum(u) = mean(C_noma_1 + C_noma_2 + C_noma_3); %Sum
capacity of NOMA

    % MIMO-NOMA

    C_mn_1 = log2(1 + pt(u)*a1.*g1./(pt(u)*a2.*g1 + no)); %User 1
    C_mn_2 = log2(1 + pt(u)*a1.*g2./(pt(u)*a2.*g1+no)); %User 2
    C_mn_3 = log2(1 + pt(u)*a2.*g2/no); %User 3
    C_mn_sum(u) = mean(C_mn_1 + C_mn_2 + C_mn_3); %Sum capacity of
NOMA

    %Sum rates
    Rn(u) = mean(R1n+R2n); %MIMO-NOMA
    Ro(u) = mean(R1o+R2o); %MIMO-OMA

    %Invidual user rates
    R1n_av(u) = mean(R1n); %MIMO-NOMA USER 1 (WEAK)
    R2n_av(u) = mean(R2n); %MIMO-NOMA USER 2 (STRONG)

    R1o_av(u) = mean(R1o); %MIMO-OMA USER 1 (WEAK)
    R2o_av(u) = mean(R2o); %MIMO-OMA USER 1 (WEAK)

```

```

% oma

C_oma_1 = (1/3)*log2(1 + pt(u)*g1/no); %User 1
C_oma_2 = (1/3)*log2(1 + pt(u)*g2/no); %User 2
%C_oma_3 = (1/3)*log2(1 + pt(u)*g3/no); %User 3

C_oma_sum(u) = mean(C_oma_1 + C_oma_2); %Sum capacity of OMA

end

```

```

figure;
plot(Pt,C_noma_sum,'-ob','linewidth',1.5); hold on;
plot(Pt,C_mn_sum,'-*r','linewidth',1.5); hold on;
plot(Pt,C_oma_sum,'-or','linewidth',1.5); hold on;
legend('NOMA ','MIMO-NOMA ','MIMO-OMA ')
xlabel('Transmit power (dBm)');
ylabel('channel capacity (bps/Hz)');
title('Channel Capacity')

```

3. Outage probability

```

%% releigh outageprobability

clc; clear variables; close all;

%Distances
d1 = 500; d2 = 200;

%Power allocation coefficients
a1 = 0.75; a2 = 0.25;

N = 5*10^5;
eta = 4;%Path loss exponent

%Rayleigh fading channels
h11 = sqrt(d1^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h12 = sqrt(d1^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h21 = sqrt(d2^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);
h22 = sqrt(d2^-eta)*(randn(N,1) + 1i*randn(N,1))/sqrt(2);

h1 = h11+h12;
h2 = h21+h22;

hf = sqrt(d1^-eta)*(randn(1,N) + 1i*randn(1,N))/sqrt(2);
hn = sqrt(d2^-eta)*(randn(1,N) + 1i*randn(1,N))/sqrt(2);
% channel gains for NOMA

```

```

gf = (abs(hf)).^2;
gn = (abs(hn)).^2;
%Channel gains
g1 = (abs(h1)).^2;
g2 = (abs(h2)).^2;

%Transmit power
Pt = -30:5:30; %in dBm
pt = (10^-3)*db2pow(Pt); %linear scale

BW = 10^6; %bandwidth
%Noise power
No = -174 + 10*log10(BW); %in dBm
no = (10^-3)*db2pow(No); %in linear scale

%Target rates
R1 = 1;
R2 = 3;

p1n = zeros(1,length(pt));%mimo-noma
p2n = zeros(1,length(pt));%mimo
p1o = zeros(1,length(pt));%mimo-oma
p2o = zeros(1,length(pt));%oma
p1 = zeros(1,length(pt));%noma
p2 = zeros(1,length(pt));%noma

for u = 1:length(pt)

    %Achievable rates for MIMO-NOMA
    R1n = log2(1 + pt(u)*a1.*g1./(pt(u)*a2.*g1 + no));
    R12n = log2(1 + pt(u)*a1.*g2./(pt(u)*a2.*g2 + no));
    R2n = log2(1 + pt(u)*a2.*g2/no);

    %Achievable rates for MIMO-OMA
    R1o = 0.5*log2(1 + pt(u)*g1/no);
    R2o = 0.5*log2(1 + pt(u)*g2/no);
    %Achivebale rates for NOMA
    R1nn = log2(1 + pt(u)*a1.*gf./(pt(u)*a2.*gf + no));
    R2nn = log2(1 + pt(u)*a2.*gn/no);
    %Sum rates
    Rn(u) = mean(R1n+R2n); %MIMO-NOMA
    Ro(u) = mean(R1o+R2o); %MIMO-OMA

    %Invidual user rates
    R1n_av(u) = mean(R1n); %MIMO-NOMA USER 1 (WEAK)
    R2n_av(u) = mean(R2n); %MIMO-NOMA USER 2 (STRONG)

    R1o_av(u) = mean(R1o); %MIMO-OMA USER 1 (WEAK)
    R2o_av(u) = mean(R2o); %MIMO-OMA USER 1 (WEAK)

```

```

R1nn_av(u) = mean(R1nn); %NOMA USER 1 (WEAK)
R2nn_av(u) = mean(R2nn); %NOMA USER 2 (STRONG)
%Outage calculation
for k = 1:N
    %MIMO-NOMA USER 1 (WEAK)
    if R1n(k) < R1
        p1n(u) = p1n(u)+1;
    end
    %MIMO-NOMA USER 2 (STRONG)
    if (R12n(k)<R1)||((R12n(k)>R1)&&(R2n(k)<R2))
        p2n(u) = p2n(u)+1;
    end
    %MIMO-OMA USER 1 (WEAK)
    if R1o(k) < R1
        p1o(u) = p1o(u)+1;
    end
    %MIMO-OMA USER 2 (STRONG)
    if R2o(k) < R2
        p2o(u) = p2o(u)+1;
    end
    % NOMA USER 1 (WEAK)
    if R1nn(k)< R1
        p1(u)= p1(u)+1;
    end
    % NOMA USER 2 (STRONG)
    if (R2nn(k)<R1)||((R2nn(k)<R2))
        p2(u) = p2(u)+1;
    end

end

end
end

figure;
semilogy(Pt,p1n/N,'-b','linewidth',1.5), hold on; grid on;
semilogy(Pt,p2n/N,'-ob','linewidth',1.5), hold on;grid on;
semilogy(Pt,p1o/N,'-r','linewidth',1.5), hold on; grid on;
semilogy(Pt,p2o/N,'-or','linewidth',1.5), hold on, grid on;
semilogy(Pt,p1/N,'-g','linewidth',1.5); hold on; grid on;
semilogy(Pt,p2/N,'-k','linewidth',1.5);
legend('MIMO-NOMA weak','MIMO-NOMA strong','MIMO-OMA weak', 'MIMO-OMA
strong','NOMA-WEAK','NOMA-STRONG');
xlabel('Transmit power (dBm)');
ylabel('Outage probability');
title('Outage comparison');

```

4.Bit Error Rate

```

clc; clear variables; close all;
N = 10^6;

```

```

d1 = 1000; d2 = 500; %Distances of users from base station (BS)
a1 = 0.75; a2 = 0.25; %Power allocation factors
eta=4; %Path loss exponent
%Generate rayleigh fading coefficient for both users
h11 = sqrt(d1^(-eta))*(randn(1,N)+1i*randn(1,N))/sqrt(2);
h12 = sqrt(d1^(-eta))*(randn(1,N)+1i*randn(1,N))/sqrt(2);
h21 = sqrt(d2^(-eta))*(randn(1,N)+1i*randn(1,N))/sqrt(2);
h22 = sqrt(d2^(-eta))*(randn(1,N)+1i*randn(1,N))/sqrt(2);
h1=h11+h12;
h2=h21+h22;
hf=h11;
hn=h22;
gf=(abs(hf)).^2;
gn=(abs(hn)).^2;
g1 = (abs(h1)).^2;
g2 = (abs(h2)).^2;
Pt = 0:2:40; %Transmit power in dBm
pt = (10^-3)*10.^(Pt/10); %Transmit power in linear scale
BW = 10^6; %System bandwidth
No = -174 + 10*log10(BW); %Noise power (dBm)
no = (10^-3)*10.^(No/10); %Noise power (linear scale)
%Generate noise samples for both users
w1 = sqrt(no)*(randn(1,N)+1i*randn(1,N))/sqrt(2);
w2 = sqrt(no)*(randn(1,N)+1i*randn(1,N))/sqrt(2);
%Generate random binary data for two users
data1 = randi([0 1],1,N); %Data bits of user 1
data2 = randi([0 1],1,N); %Data bits of user 2
%Do BPSK modulation of data
x1 = 2*data1 - 1;
x2 = 2*data2 - 1;
p = length(Pt);
for u = 1:p
%Do superposition coding
x = sqrt(pt(u))*(sqrt(a1)*x1 + sqrt(a2)*x2);
%Received signals
y1 = h1.*x + w1;
y2 = h2.*x + w2;

yf=hf.*x+w1;
yn=hn.*x+w2;

%Equalize
eq1 = y1./h1;
eq2 = y2./h2;

eq3=yf./hf;
eq4=yn./hn;

%AT USER 1-----
%Direct decoding of x1 from y1

```

```

x1_hat = zeros(1,N);
x1_hat(eq1>0) = 1;
%Compare decoded x1_hat with data1 to estimate BER
ber1(u) = biterr(data1,x1_hat)/N;

xf_hat = zeros(1,N);
xf_hat(eq3>0) = 1;
%Compare decoded x1_hat with data1 to estimate BER
ber1f(u) = biterr(data1,xf_hat)/N;
%-----
%AT USER 2-----
%Direct decoding of x1 from y2
x12_hat = ones(1,N);
x12_hat(eq2<0) = -1;
y2_dash = eq2 - sqrt(a1*pt(u))*x12_hat;

x1n_hat = ones(1,N);
x1n_hat(eq4<0) = -1;
yn_dash = eq4 - sqrt(a1*pt(u))*x1n_hat;

x2_hat = zeros(1,N);
x2_hat(real(y2_dash)>0) = 1;
ber2(u) = biterr(x2_hat, data2)/N;

xn_hat = zeros(1,N);
xn_hat(real(yn_dash)>0) = 1;
ber2n(u) = biterr(xn_hat, data2)/N;

end
semilogy(Pt, ber1,'r','linewidth',1.5); hold on; grid on;
semilogy(Pt, ber2,'b','linewidth',1.5);
semilogy(Pt, ber1f,'k','linewidth',1.5); hold on; grid on;
semilogy(Pt, ber2n,'g','linewidth',1.5);
xlabel('Transmit power (P in dBm)');
ylabel('BER');
title('BIT ERROR RATE');
legend('User 1/Far user MIMO-NOMA','User 2/Near user MIMO-NOMA','User 1/Faruser NOMA','User 2/Near user NOMA');

```

B. Screen Shots

